

ENEE 381 Problem Set #4

10/14/04 due 10/26/04

Questions like (1) -(6) and (9) -(14) could easily be on the first examination

- (1) A length of loss-less transmission line is first short-circuited at one end and then open-circuited. The impedance measured at the other end in the first case is Z_1 and Z_2 in the second. Prove that $Z_1 Z_2 = Z_0^2$. This is a convenient way for measuring the characteristic impedance of an unknown line.
- (2) A 50 ohm transmission line is terminated with an impedance of $20-j30$. What is the magnitude and phase of the reflection coefficient?
- (3) Repeat question (2) using the Smith chart
- (4) A 75 ohm transmission line is terminated with a load of $150 + j50$ ohm. Compute ρ in terms of both amplitude $|\rho|$ and ϕ . What fraction of incident power is absorbed in the load?
- (5) Use Smith chart. A line with $Z_0 = 100\Omega$ is terminated with an unknown load. The SWR is found to be 3. A current maximum is observed 0.1λ from the load. What are:
 - (a) the load?
 - (b) the reflection coefficient ρ , magnitude and angle?
 - (c) how would you match the line without changing the load at the end of the line?
- (6) Use Smith chart. A transmission line of characteristic impedance 75 ohm is terminated with an impedance $50+j125$ ohm. 0.1λ from the load a 150ohm shorted stub 0.2λ long is connected in shunt to the main line. What are:
 - (a) The reflection coefficient in magnitude and phase at this point?
 - (b) The standing wave ratio?
 - (c) Where is the nearest current minimum that is greater than 0.1λ from the load?
 - (d) Where is the nearest point greater than 0.1λ from the load where the line can be matched with an open 75 ohm stub?
- (7) RWvD 5.2b
- (8) RWvD 5.5b
- (9) RWvD 5.7f Do with and without the Smith Chart
- (10) RWvD 5.8a Use Smith Chart
- (11) RWvD 5.8c Use Smith Chart
- (12) RWvD 5.10a Use Smith Chart
- (13) RWvD 5.10b Use Smith Chart
- (14) RWvD 5.10g Use Smith Chart

ENEE 381 Problem Set 4. SOLUTIONS

(1) The transformed impedance equation is

$$Z_i = Z_0 \frac{(Z_L \cdot \cos(kl) + j \cdot Z_0 \cdot \sin(kl))}{Z_0 \cdot \cos(kl) + j \cdot Z_L \cdot \sin(kl)}$$

For a shorted line $Z_L = 0$, so

$$Z_{\text{short}} = j \cdot Z_0 \cdot \tan(kl)$$

For an open line Z_L is infinite, so

$$Z_{\text{open}} = -j \cdot Z_0 \cdot \cot(kl)$$

$$Z_{\text{short}} Z_{\text{open}} = Z_0^2 \quad \text{Q.E.D.}$$

(2) $Z_0 := 50$ $j := i$

$$Z_L := 20 - j \cdot 30$$

$$\rho := \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$\rho = -0.207 - 0.517i$$

$$|\rho| = 0.557 \quad \text{magnitude}$$

$$\frac{\arg(\rho)}{\text{deg}} = -111.801 \quad \text{phase angle in degrees}$$

(3) On chart the normalized impedance is

$$\zeta_L := \frac{Z_L}{Z_0}$$

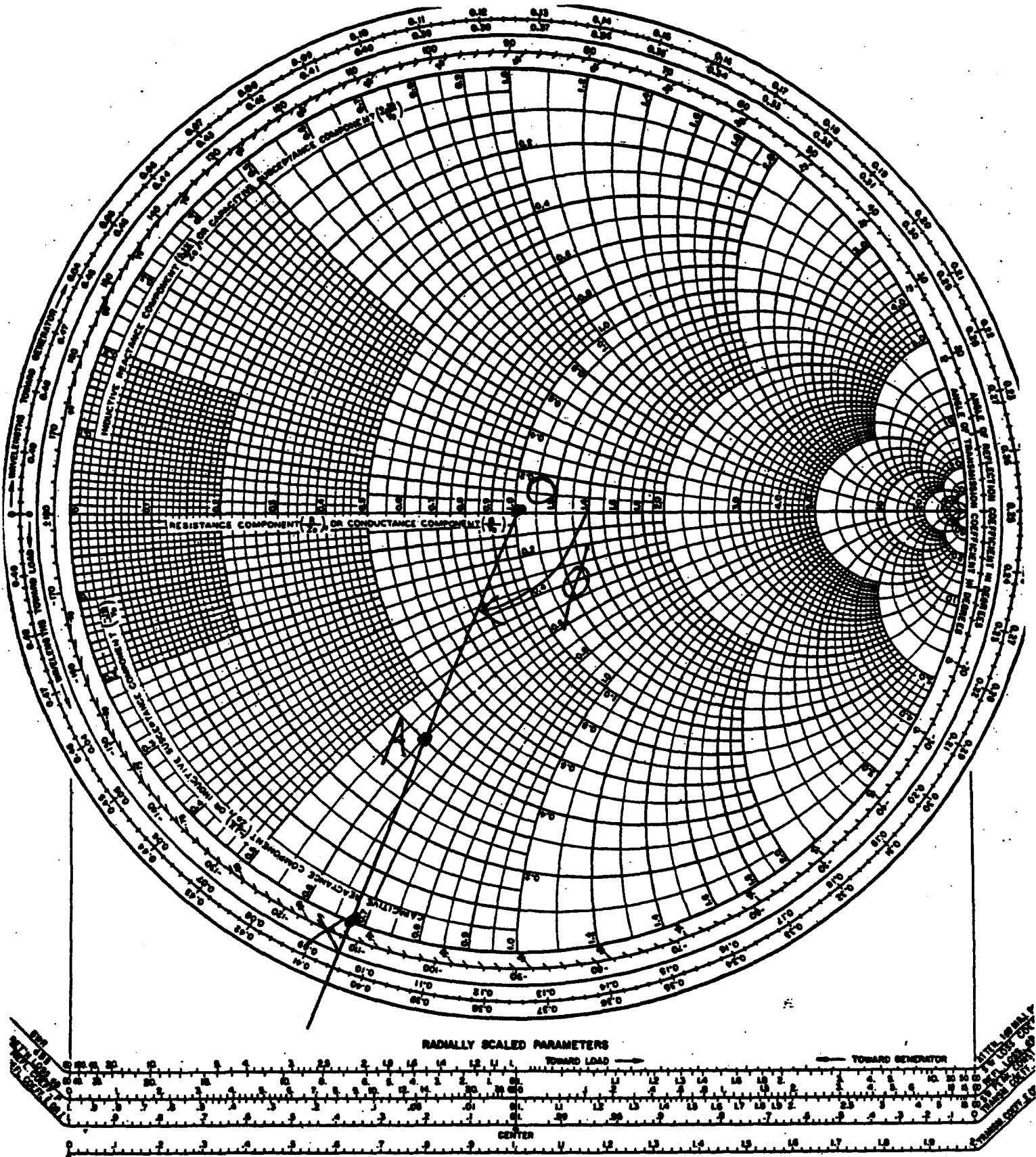
$$\zeta_L = 0.4 - 0.6i \quad \text{Point A on chart}$$

(4) $Z_0 := 75$ $Z_L := 150 + j \cdot 50$

$$\rho := \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$(|\rho|)^2 = 0.153 \quad \text{Fraction of power reflected}$$

$$1 - (|\rho|)^2 = 0.847 \quad \text{Fraction of power absorbed in load}$$



$$(3) \quad Z_L = 0.4 - j0.6$$

$$|r| = \frac{OA}{OX} = 0.56 \quad \phi = -111^\circ$$

(5)

$$Z_0 := 100 \quad S := 3$$

On chart mark point on real axis for $S=3$, draw circle about center of chart Γ_{\max} is on left hand side. Load is 0.1λ away at point $0.5-j0.6$

$$|\rho| = 0.49 \quad \phi = 108 \text{ degrees}$$

Sanity check

$$\zeta_L := 0.5 - j0.6$$

$$Z_L := Z_0 \cdot \zeta_L$$

$$\rho := \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$|\rho| = 0.483 \quad \frac{\arg(\rho)}{\text{deg}} = -108.004$$

Matching points are at B or C. You can match with series or parallel components but the distances from the load vary depending on whether you use the chart in impedance or admittance

$$Z_0 := 75 \quad Z_L := 50 + j125$$

$$\zeta_L := \frac{Z_L}{Z_0} \quad \text{Normalized load}$$

$$\zeta_L = 0.667 + 1.667i \quad \text{point A}$$

$$y_L := \frac{1}{\zeta_L}$$

$$y_L = 0.207 - 0.517i \quad \text{Normalized admittance of load, point B}$$

Shorted stub 0.2λ long starts at point C and has its normalized admittance at point D

$$y_{\text{stub}150} := -j0.32$$

$$\text{Renormalized } Y_{\text{stub}} := \frac{1}{150} \cdot y_{\text{stub}150} \quad \text{multiplying by admittance of 150 ohm line}$$

Normalize again but this time to 75 ohm line

$$y_{\text{stub}75} := \frac{Y_{\text{stub}}}{\frac{1}{75}}$$

$$y_{\text{stub}75} = -0.16i$$

(6)

0.1λ from the load is point E, where

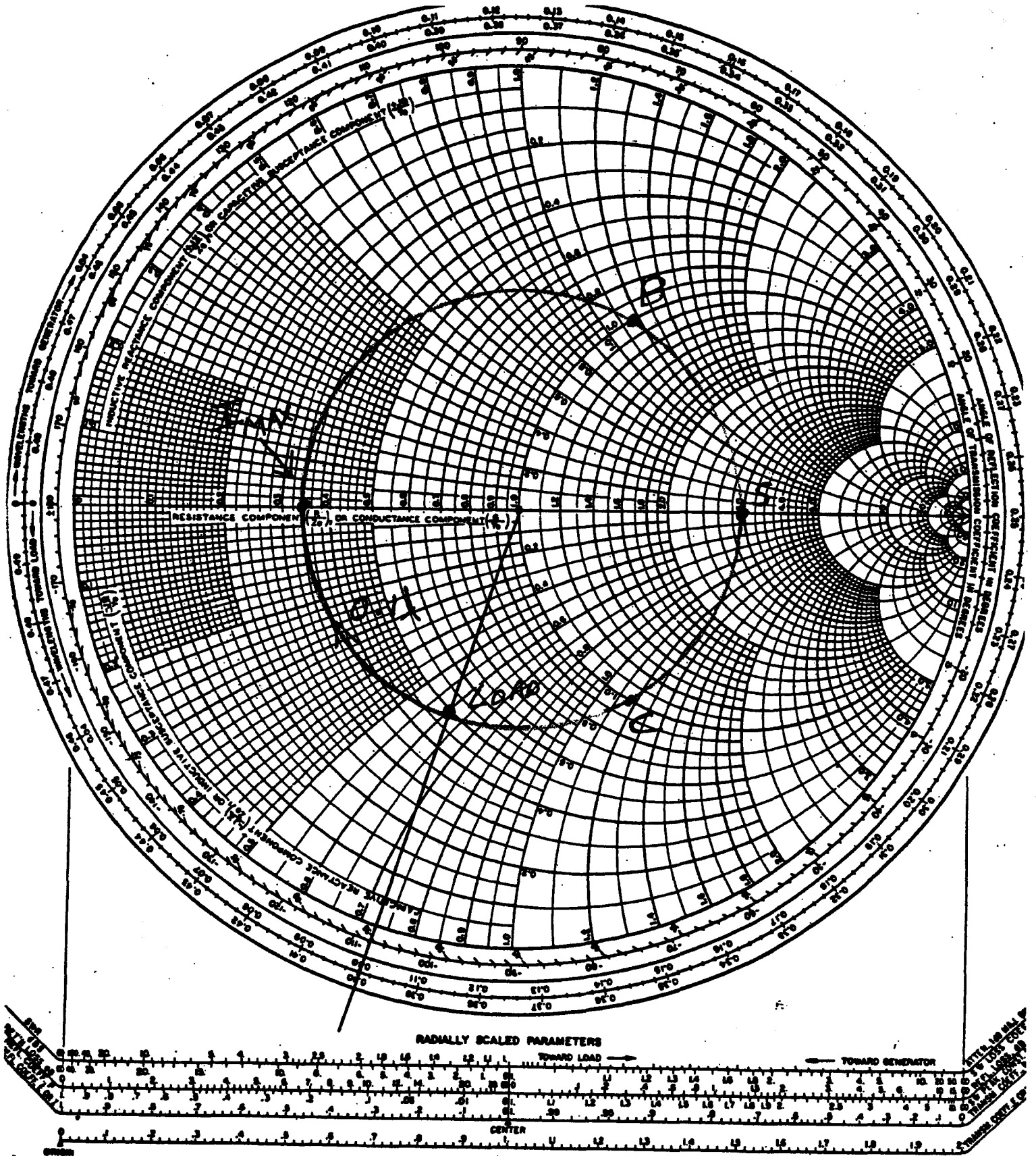
$$y_i := 0.17 + j-0.14$$

After adding stub

$$y_{total} := y_i + y_{stub75}$$

$$y_{total} = 0.17 - 0.02i \quad \text{Point F}$$

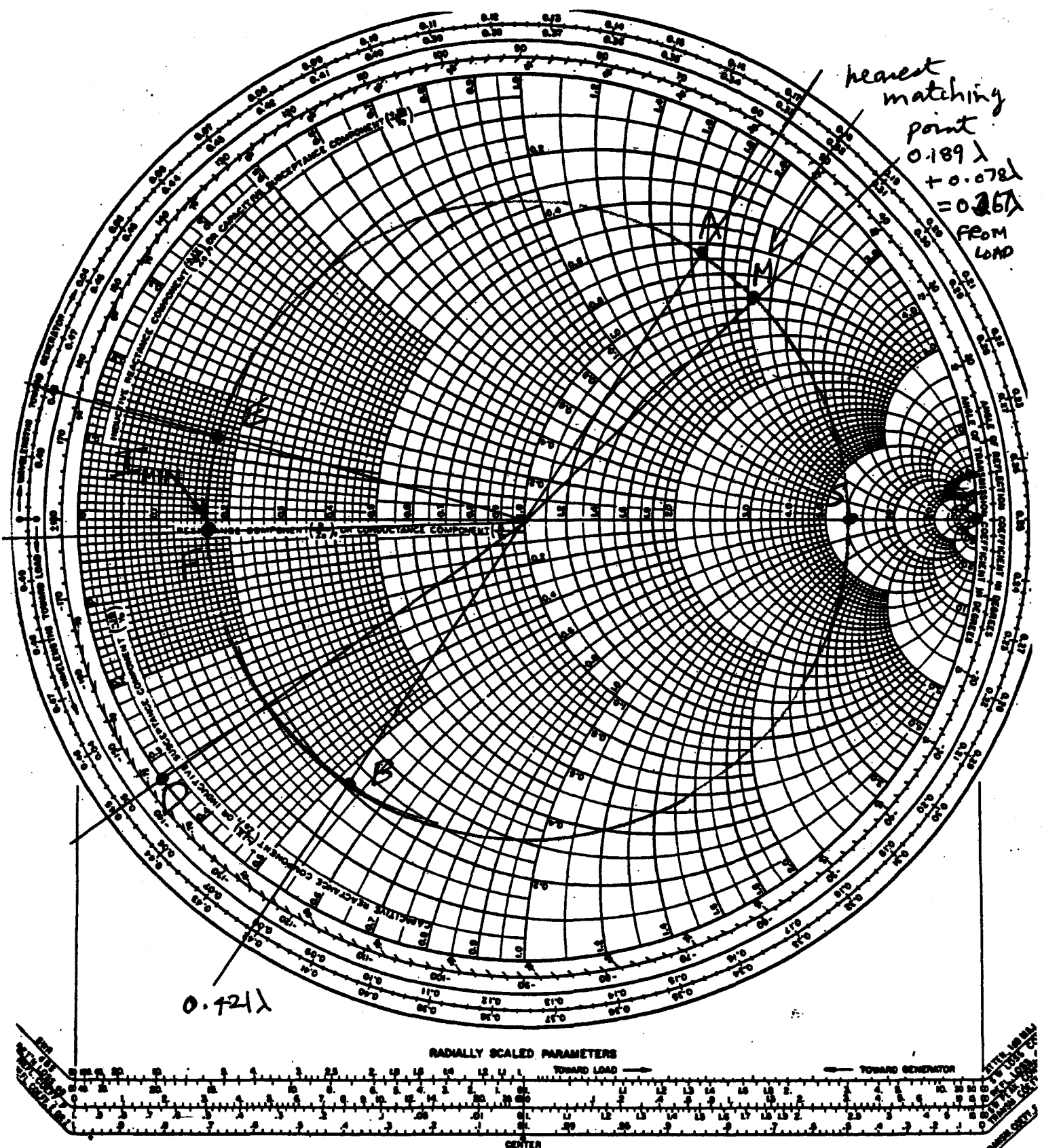
See chart for rest of problem



(5) $\Gamma_L = 0.5 - j0.6$ $Z_L = 50 - j60$

$|\Gamma| = 0.99$ $\phi = -108^\circ$ MATCH AT B OR C

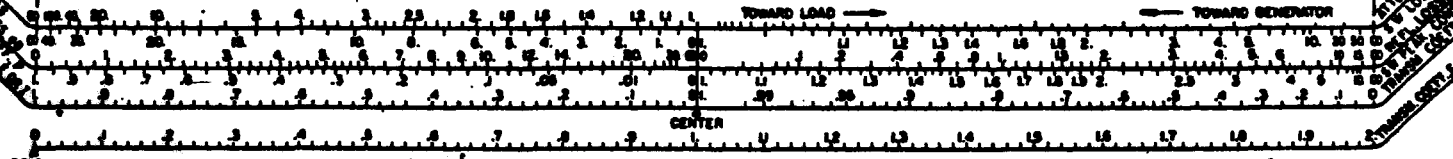
EXAMPLE B = $1 + j1.16$ ADD normalized
 $-j1.16$ IN SERIES



nearest
matching
point
 0.189λ
 $+ 0.072\lambda$
 $= 0.261\lambda$
FROM
LOAD

0.72λ

RADIALLY SCALED PARAMETERS



(6)

At F $|e| = 0.71$ $\phi = -178^\circ$ $S = 6$
 S is also the location of I_{MAX} , I_{MIN} is
 just past point F. The nearest I_{MIN}
 $> 0.1\lambda$ from load is $0.078\lambda + 0.5\lambda = 0.578\lambda$
 from load

(7)

5.2b: From eqs 1.9(3) and 2.5(3), for parallel plane line

(i) $C = \frac{\epsilon W}{a}$ F/m and $L = \frac{\mu a}{W}$ H/m, so $Z_0 = \frac{a}{W} \sqrt{\frac{\mu}{\epsilon}}$, $v = \frac{1}{\sqrt{\mu \epsilon}}$

(ii) For $W = 5 \mu\text{m}$, $a = 1 \mu\text{m}$, $\mu = \mu_0$, $\epsilon = 2.5 \epsilon_0$

$$Z_0 = \frac{1}{5} \sqrt{\frac{4\pi \times 10^{-7}}{2.5 \times 8.854 \times 10^{-12}}} = 47.7 \Omega, \quad v = \frac{c}{\sqrt{2.5}} \approx 1.9 \times 10^8 \text{ m/sec.}$$

(iii) For $a = .5 \mu\text{m}$, other values as above,

$$Z_0 = \text{half of above} = 23.8 \Omega, \quad v = \text{unchanged in this model.}$$

(8)

5.5b: Time to propagate 200 m,

$$t_1 = \frac{200}{2 \times 10^8} = 10^{-6} \text{ sec} = 1 \mu\text{sec.}$$

So reflection occurs after 1 μsec

$$\frac{V_-}{V_+} = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{Z_{22} - Z_{01}}{Z_{22} + Z_{01}} = \frac{50}{150} = \frac{1}{3}$$

but $\frac{I_-}{I_+} = -\frac{1}{3}$, $I_+ = \frac{V_+}{Z_0} = \frac{100}{50} = 2 \text{ A}$

Note current continuous at $z = 200$

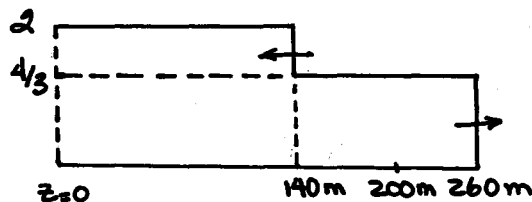
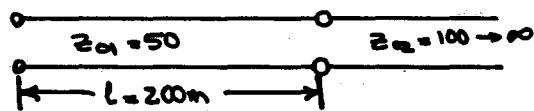
after reflection. So plot at

$t = 1.3 \mu\text{sec}$ to right

$$W_{T_+} = V_+ I_+ = 100 \times 2 = 200 \text{ W}$$

$$W_{T_-} = V_- I_- = \frac{100}{3} \times \frac{2}{3} = \frac{200}{9} \text{ W}$$

$$(W_T)_2 = V_2 I_2 = \frac{200}{150} \times 100 \times \frac{4}{3} = \frac{1600}{9} \text{ W}$$



$$\left. \begin{array}{l} \text{but } 200 - \frac{200}{9} = \frac{1600}{9} \\ \text{So there is a power balance.} \end{array} \right\}$$

(9)

5.7f: $Y_L = .05 + j2\pi \times 10^{10} \times 10^{-12} = .05 + j.0628 \text{ S}$, $Y_0 = .05 \text{ S}$,

$$\rho = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{Y_0 - Y_L}{Y_0 + Y_L} = \frac{-j.0628}{.10 + j.0628} = -.50 - j.47 = .56 \angle 237.4^\circ$$

$$Y_i = Y_0 \left[\frac{Y_L \cos \beta L + j Y_0 \sin \beta L}{Y_0 \cos \beta L + j Y_L \sin \beta L} \right] = .05 \left[\frac{(5 + j6.28) \sqrt{2}/2 + j5 \times \sqrt{2}/2}{5 \sqrt{2}/2 + j(5 + 6.28) \times \sqrt{2}/2} \right]$$

$$Y_i = (-9.42 + j7.44) 10^{-2} = .12 \angle 141.7^\circ$$

(10)

5.8a: $\rho = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{(100 + j100) - 50}{(100 + j100) + 50} = \frac{7 + j47}{15} = .62 \angle 29.7^\circ$

$$S = \frac{1 + |\rho|^2}{1 - |\rho|^2} = \frac{1 + .62}{1 - .62} = 4.27$$

(11)

5.8C: This problem is readily done on the Smith chart, but numerical solution given here:

$$|p| = \frac{S-1}{S+1} = \frac{1.5-1}{1.5+1} = .2$$

$$\beta z_{\max} = \beta z_{\min} + \frac{\pi}{2} = -\frac{2\pi}{\lambda} (.1\lambda) + \frac{\pi}{2} = .3\pi$$

$$\text{so } \varphi = -2(\beta z)_{\max} = -.6\pi = -108^\circ, \text{ so } \rho = .2 \angle -108^\circ$$

$$z_i = z_0 \left[\frac{1 + \rho e^{-2j\beta L}}{1 - \rho e^{-2j\beta L}} \right] = z_0 \frac{1 + .2e^{-j(6+.8)\pi}}{1 - .2e^{-j(6+.8)\pi}}$$

$$z_i = z_0 \left[\frac{1 + .2e^{j.6\pi}}{1 - .2e^{j.6\pi}} \right] = [.827 + j.328] z_0$$

(12)

5.10a: $\frac{z_L}{z_0} = \frac{75 - j369}{50} = 1.50 - j7.38$ point A; Smith chart on next page.

$$\frac{l}{\lambda} = \frac{35 \times 50 \times 10^6}{3 \times 10^8} = .583$$

we subtract a half wave and add .083 to .3145 getting .3975.

$$\text{Read } \frac{z_i}{z_0} = .46 - j.65$$

$$z_i = (23 - j32.5) \Omega$$

SWR - read from C:

$$S = 3.3$$

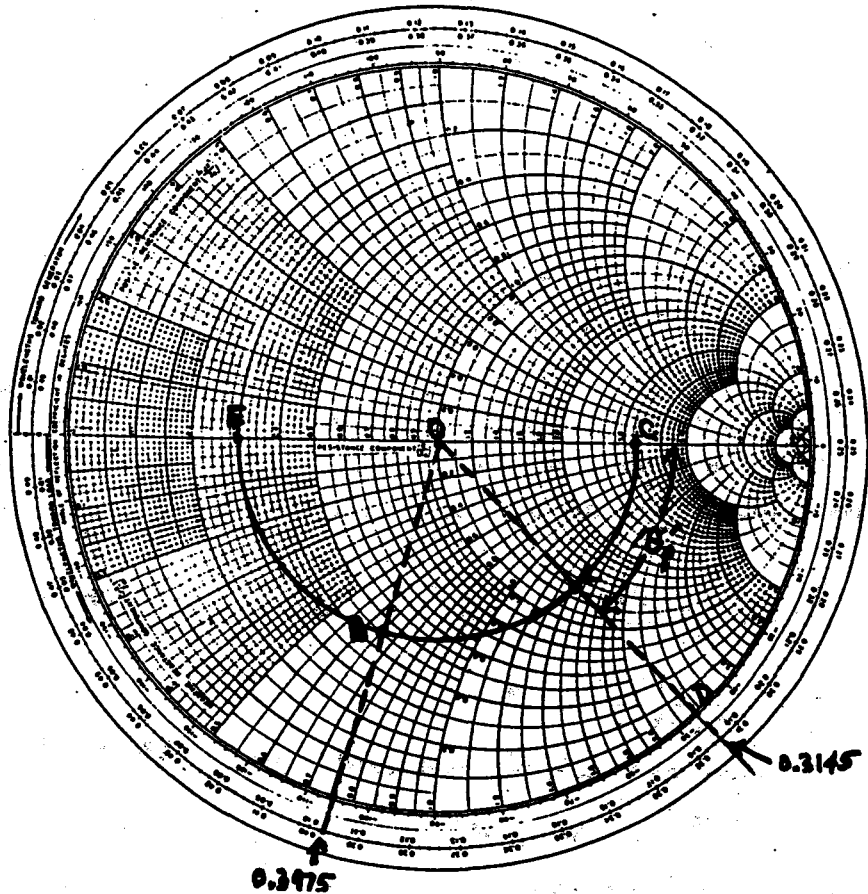
$$|p| = \frac{S-1}{S+1} = \frac{2.3}{5.3} = .55$$

$$\text{check } |p| = \frac{S-1}{S+1} = \frac{2.3}{5.3} = .54 \text{ (fair)}$$

$$\varphi = - \left(\frac{.3145 - .25}{.5} \right) \times 360^\circ = -46.4^\circ \quad (\text{continue.})$$

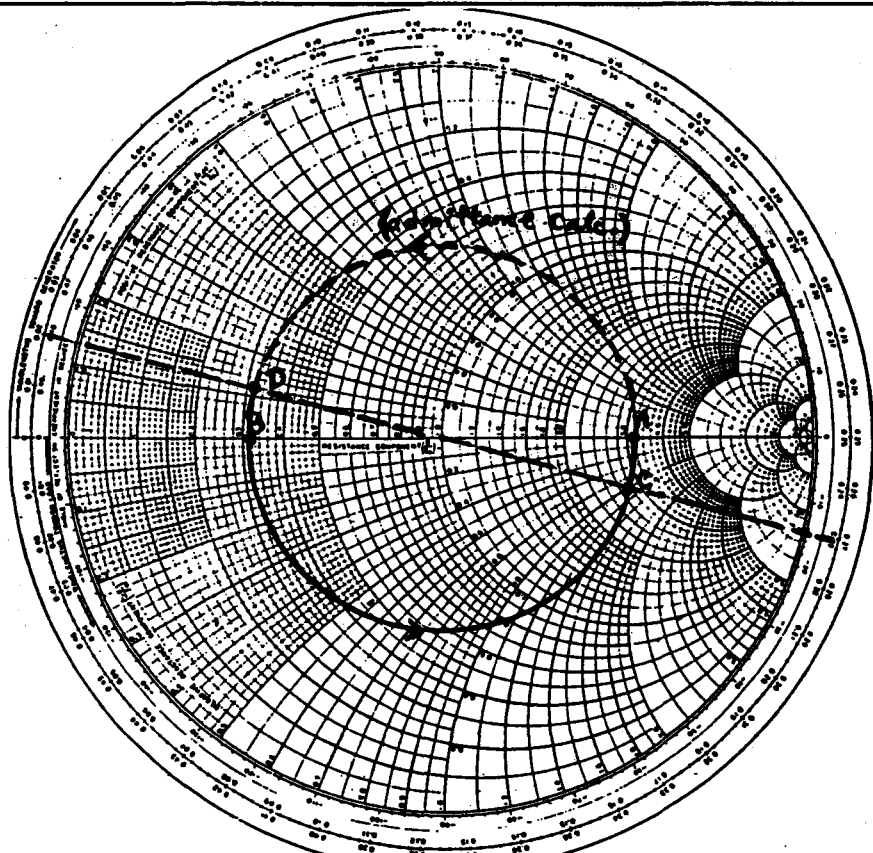
(12) continued

Position of voltage minimum is at $E, (.5 - .3145)\lambda$
 From load or 1.11 meters in front of load. (check from equations gives slightly different numbers because of inaccuracies in reading chart.)



5.10b: Enter $S = 3.2$ at A
 voltage minimum, at
 same radius, pt. B
 Then move .23 λ from
 B toward load to get
 load impedance, pt. C
 Read $\frac{Z_L}{Z_0} = 2.7 - j1.0$
 or $Z_L = 70(2.7 - j1.0)$
 $= 189 - j70$

(13)



(14)

5.10g: Since shorted stubs are in parallel with main line, we use admittances.

$$Y_L = \frac{50}{20 + j30} = .77 - j1.15 \text{ (point A of chart)}$$

Moving $.2\lambda$ toward generator we reach point B, $.30 + j.27$, the shorted stub can add susceptance only. We

want to add enough so that when transformed $\lambda/4$ we end on the circle $1 + jb$.

Then stub 2 can be adjusted to cancel b and provide a match. It is thus

helpful to transform the circle K (locus of $1 + jb$) $\lambda/4$ to give circle K' , the

transformed locus. Then stub 1 should add enough susceptance to go from

B to C on K' , addition of $Y_1 = j.20$ gives C, $.30 + j.47$. Moving $\lambda/4$ toward

load then gives D, $1.0 - j1.5$ and addition of $Y_2 = j1.5$ gives match, 0. Then

length L_1 of stub 1, $-jY_{01} \cot \beta L_1 = jY_0 (.20)$, $Y_0 = Y_{01} = 1/50$, so $\beta L_1 = 101.3^\circ \Rightarrow L_1 = .281 \lambda$.

For stub 2, $-jY_{02} \cot \beta L_2 = jY_0 (.15)$, $Y_0 = Y_{02} = 1/50 \Rightarrow \beta L_2 = 146.3^\circ \Rightarrow L_2 = .406 \lambda$. (alternatively

could have added negative susceptance to get to point C' - then L_1 & L_2 each shorter than above by $\lambda/4$.)

